

epc700/epc702

24V/50mA General-Purpose Output-Driver





Absolute Maximum Rat	ings (Note 1)	Recommended Ope	rating Co	onditions	
			Min.	Max.	Units
Power Supply Voltage V_DD	-0.3 to +36.0 V (Note 2)	Power Supply Voltage (VDD)	9.6	30	V
Maximum Power Dissipation	100mW				
Storage Temperature Range (Ts)	-40°C to +85°C	Operating Temperature (T _o)	-40°	+85°	С
Lead Temperature solder, 4 sec. (T_L)	+260°C	Humidity (non-condensing)	+5	+95	%

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Recommended operating conditions indicate conditions for which the device is intended to be functional, but do not guarantee specific performance limits. For guaranteed specifications and test conditions, see Electrical Characteristics.

Note 2: Supply voltages up to 36 Volts may be present for 10 seconds only.

Note 3: This device is a highly sensitive CMOS amplifier with an ESD rating of JEDEC HBM class 1C (>1kV). Handling and assembly of this device should only be done at ESD protected workstations.

Electrical Characteristics

 V_{DD} = 9.6V < V_{DD} < 30V, -40°C < T_A < +85°C

Symbol	Parameter	Conditions/Comment					
				Min.	Тур.	Max.	Units
V _{DD}	Supply Voltage			9.6		30	V
ΔV_{DD}	Ripple on Supply Voltage	Peak-Peak				10	%V _{DD}
I _{DD}	Supply Current	no load			300	400	μA
Vout	Output Voltage		0		30	V	
V_{Sat}	Output Saturation Voltage	@50mA output current			1	2	V
I _{SENS}	Sens Current	Current trigger threshold		-50	-60	-70	mA
V_{SENS}	Current Sens Voltage	Over-current trigger threshold voltage (by power switch)	0.18	0.2	0.25	V	
I _{Peak}	Short Circuit Peak Current	Initial current during a short circuit (<1ms,			-0.5	A	
V _{Status} Status Output		Logical high	2		5.5	v	
		Logical low	-0.3		0.8	1 ×	
		Sink driving capability		-8	-10	-12	mA
\mathbf{f}_{Status}	Status Output Frequency	epc702 only, duty cycle 50%	1.5	1.7	1.9	Hz	
V _{IN}	Input	Logical high		2.0		5.5	
		Logical low		-0.3		0.8	V
		Hysteresis		0.25			
		Pull-down resistance	100	150	200	kΩ	
P _{DIS}	Power Dissipation	On-chip power dissipation				100	mW
t _{on}	Response Time	On			1.0	1.2	μs
t _{OFF}	Response Time	Off			0.7	1.0	μs
t _{del}	Off-delay Time	Time between over-current detection and	epc700-CSP6-A	40	50	60	
		STATUS/OUT change (default value), re- fer to section Programming	epc702-CSP6 epc700-QFN16 epc702-QFN16	80	100	120	μs
		Programmable off-delay values	5/10/20/5	50/100/200	/500/1,000		
\mathbf{t}_{minOFF}	Recovery Time	Minimum down time of the OUT pin to	epc700-CSP6-A	400	500	600	ms
		protect the external transistor (default value), refer to section Programming	epc702-CSP6 epc700-QFN16 epc700-QFN16	18	20	22	
		Programmable values	10/20/50/100/200/500/1,000/ 1,500				



Sym-	Pa	rameter		Condi	litions/C	Commen	its			Values			
bol									Min.	Тур.	Max.	Units	
f _{rt}	Short circu time factor	iit recovery delay	t _{minoff} /t _{delay} = fer to sect	f _{rt} (when used ion "Over-curre	without ent Rese	external t Sequer	driver tran	sistor, re-	1,000				
t _{startup}	Start-up tir	ne	VDD ramp	o > 100 V/ms							200	μs	
C_{L_max}	External Lo	oad Capacitance	in be dri 2kOhm	ven throu load and	ugh OUT w 1 5µs delay	ithout time		30		nF			
	C	onnection I	_	ns				71	LL 0L	6			
		OUT SENS) (4	US		OUT SENS V _{DD}	13 14 15 16 O		Image: Contract of the second secon	6	AI 4	ND I TATUS	
	6-1	Pin Chip Scale P	ackage (CS	SP)		No	te: For san	16-Pi	n QFN Pa	ckage	Please ing	iiro	
6-Pin CSP	16-Pin QFN	Pin Nam	e	Description		Note: For sampling only. Limited quantities. Please inquire.							
1	8	GND		Negative powe	er suppl	y pin							
2	7	IN		Input from the	controll	er							
3	5	STATUS	6	Output status	signal to	I to the controller							
4	16	V _{DD}		Positive power	r supply	ply pin							
5	14	SENS				ch internal/external mode or to measure the voltage drop on an external etect over-current							
6	13	OUT		Output									
n/a	1-4, 6, 9-12, 15	NC		Not connected	d. Conne	ect these	pins to GN	ID.					



Functional Description

Normal Operation

During typical operation the OUT follows the IN as shown in Figure 1. As long as the output current does not exceed the current threshold, OUT is stable ON and STATUS is stable HIGH-Z. The delay from IN to OUT is defined as t_{ON} and t_{OFF} for the rising and the falling edge respectively.

IN									
OUT (without external Transistor)	HIGH-Z Vsat								
OUT (with external Transistor)									
SHORT									
SENS	0								
STATUS epc700									
90102	Figure 1: Normal operation								
Please note that the status of the pin OU quent diagrams, the version with external	T is dependent whether the chip is operated with or without external driver transistor. In the subse- driver transistor is shown only.								
Over-current Sequence									
A short on the load side will lead to an over-current through OUT. If an over-current stays longer than the time t_{del} and t_{ext} , OUT is turned off as shown in the figure below. At the same time STATUS changes its state to indicate an over-current situation to an external controller. STATUS can also be used to drive directly an indicator LED due to its 10mA driving capability. epc700 delivers a constant on-signal, whereas epc702 has a flashing output.									
OUT 0									
SHORT 0									
SENS 0	V _{th sens}								
	text del								
STATUS ^{HIGH-Z} epc700									
STATUS HIGH-Z epc702									
	Figure 2: Short circuit detection								
Note: The parameter t_{ext} used in this paper	r is described in Figure 7.								

Current Peak at OUT

A short current peak when OUT is turned on, typically generated by a capacitive load, could trigger the short-circuit protection logic. However, if the current peak is shorter than t_{ext} plus t_{del} , the over-current peak will be ignored.

Make sure that the energy drawn by such a current peak does not destroy the internal/external output transistor.



Figure 3: Short over-current pulse, i.e., by switching a capacitive load

Over-current Reset Sequence

If there is a permanent short circuit at OUT, such short circuit will be detected and OUT is turned off. After a waiting time t_{minOFF}, the device tries to turn on OUT again. If the short circuit is still present, OUT is immediately turned off again. This sequence continues until the short circuit is removed or IN goes to LOW or power is turned off. This mode is called self-healing since the device tries to self-heal the short circuit end to switch back into normal operation.

As a consequence in the case of a permanent over-current, short current peaks are issued into the load, respectively short.

The time t_{minOFF} has to be set to a value that the internal/external switch cannot be damaged by a too high power dissipation. Without an external switch, the time t_{minOFF} has to be 1,000 times longer than the time t_{del} .



Application Information

epc700 and epc702 have two modes of operation, where the SENS pin is used to define the mode. When SENS is tied to VDD, the chip operates as a sink driver capable to sink max. 50mA at 30VDC (refer to Figure 6). The load is connected directly between V_L and the OUT pin. If the current through the internal switch exceeds I_{SENS} , the switch is turned off.

If the SENS pin is at low level, the OUT pin is driven by a source driver also capable to drive 50mA into an external power transistor. This mode is used if the required output current has to be higher than 50mA, e.g. 1A and the output voltage exceeds 30VDC (refer to Figure 7).

The load current is measured by monitoring the voltage drop over a resistor. If the internal switch is used, also the current measurement resistor is located internally (Figure 6). In the case of using an external power transistor as shown in Figure 7, the current measurement resistor R_{SENS} has to be placed externally. If the voltage drop at R_{SENS} exceeds the threshold of 200 mV, the output stage is deactivated. The timing diagrams of the signals can be found in section "Functional Description".

The IN signal must be low during power-up (t_{STARTUP}) for proper function of the chip. The epc70x has a built in pull down resistor, so not external active driving is needed during startup.

epc700 or epc702 Using the Internal Switch

Figure 6 shows the epc700/702 in the mode using the internal switch. To enable this mode, the SENS pin has to be connected to VDD. Note that the VDD of the chip and V_L at the load can have a different value. The values for both VDD and V_L need to be between 9.6 and 30V.

The factor f_{rt} between minimum off-time and delay-time must be maintained in order not to damage the chip due to overheating. This factor has to be higher than 1,000. In the worst case scenario a peak current of approx. 0.5A is flowing from V_L at 30V into the chip with a t_{Del} set at e.g. 50µs if a short-circuit occurs. If the recovery time t_{minoff} in this case is smaller then e.g. 50ms, the average power dissipation would exceed the safe operation condition and the device will get damaged.

The diode D_{L} is to protect the internal switch against voltage surges when inductive loads are turned off.

R1 is to protect the internal switch in case of a short circuit on the load when a very low impedance power supply is used.

The voltage V_L can be higher than VDD in this configuration. However, it must not be above the maximum value of 'Supply Voltage' stated in the table Electrical Characteristics.

epc700 or epc702 Using an External Switching Transistor

Figure 7 shows the operation mode using an external switch T1 in order to extend the output current/voltage drive capability. In this example, a bipolar transistor is used, whereas the base current is limited by the resistor R1. The maximum base current is 50mA. In order not to damage the chip, the user has to select the resistor R1 such that the chip does not need to drive more than 50mA. Possible switches are a NPN BJT or an n-channel MOSFET.

The load is turned on and off by setting the pin IN to high respectively to low level. When the load is turned on, the load current flows from V_L through the resistor R_{SENS} and through the transistor T1 to GND. This current creates a voltage drop over R_{SENS}. The resulting voltage is applied to pin SENS, which measures the voltage drop. If it exceeds the threshold of an internal comparator, set to 200mV, the output is turned off after the given delay time t_{del}.

If the delay time should be extended to a value above the possible settings of t_{del} (refer to Table), an RC network can added, designated as R_T and C_T in Figure 7. The additional time delay t_{ext} can be calculated approx. as $R_T \times C_T$. However, the time varies according to the current through R_{SENS} . This design concept is especially useful, when a large capacitor in the load path needs to be charged. The additional delay in the over-current detection helps in such a situation.

Note that the VDD of the chip and V_L on the load are different in most of the applications. The value of VDD must be between 9.6 and 30V. V_L instead, can be on a level which is appropriate to the external switching transistor.

The diode DL is to protect the transistor T1 against voltage surges when inductive loads are turned off.



Figure 6: epc700 or epc702 using the internal switch to drive a load of up to 50mA /30VDC



Figure 7: epc700 or epc702 operation mode using an external switching transistor. In case of a short-circuit in the load the turn-off delay can be extended by an external RC network. This network adds t_{ext} to the internal delay t_{det}.



Programming

The time delay (t_{del}), until the output is turned off after the detection of an over-current condition can be programmed in order to adapt the timing to specific requirements, i.e., if a capacitive load has to be operated or an external transistor allows other values. The default value is e.g. 50µs for the epc700-CSP6-A which allows to charge a load capacitor of approx. 100 – 500nF without an external power transistor, dependent on the source impedance, the load impedance and the voltage V_L.

The time until the output is turned on again after a short circuit can be programmed as well (t_{minOFF}). This "self-healing" mechanism is very useful because no operator interaction is necessary after a short circuit to enable normal function once the short circuit has been eliminated. The default value is e.g. 500ms for the epc700-CSP6-A which means that the device tries to turn the output on after 0.5s waiting time in the short circuit mode. This waiting time is recommended as long as t_{del} is not changed. If t_{del} has been changed, the parameter t_{minOFF} shall be changed accordingly in order to respect parameter f_{rt} . The user has to ensure that the maximum operation conditions never exceed in order to avoid damage of the device.

It is to note that the parameters programmed are stored in a volatile random access memory. Thus, the parameters will be lost after a power down for longer than 5ms (data retention time @ 25°C: min. 100ms). The corresponding requirements for safety applications have to take in consideration. Parameters can be changed as many times as necessary and even under operation to change the behavior of an output. During power-on, the default values are restored automatically.

Programming Interface

The interface to store changed parameters are the pins IN and STATUS. IN is the chip select pin and STATUS, which is under normal operation an output, is used as an input pin. As long as the IN pin is at low state, parameters can be stored through the STATUS pin. Since IN is low during the programming of new parameters, OUT is low as well.

The digital input high threshold is typically at 2.2V, thus a 5V compatible communication. Please note that the voltage at STATUS should not exceed 5.5V.

Single Wire Communication Interface

The epc70x is based on a single wire communication interface by using the STATUS pin. Programming is done by a 21-Bit Manchester code according to IEEE 802.4.



Figure 8: Manchester encoding sample

Figure 8 shows such a sample Manchester encoded data-stream. The clock and the corresponding data is used to generate the Manchester data-stream. Each positive clock-edge in the Manchester encoded data (indicated with the up-arrow) corresponds to a 1 and each negative clock-edge (indicated with the down-arrow) corresponds to a 0.

Data Clock Frequency Range

The communication frequency range has to be according to Table 1.

	minimal	typical	maximum
Data clock	396kHz	450kHz	540kHz

Configuration Bit Stream for changing the delay time

In order to guarantee a reliable communication with the Manchester encoded bit stream on STATUS, some additional bits have been added to the configuration bits. Table 2 shows the digital pattern for the delay time configuration and the recovery time configuration.

Bit #	0	1	2	3	4	5	6	7	8	9	10	11	12	13	14	15	16	17	18	19	20
t _{del} Value	1	0	1	1	0	0	0	0	0	1	1	1	1	0	D1	D2	D3	1	d1	d2	d3
t _{minOFF} Value	1	0	1	1	0	0	0	0	1	0	0	0	1	0	01	02	03	1	01	o2	o3

Table 2: Configuration of the delay and the recovery time





The delay time is set with d_1 , d_2 , and d_3 . The bits D_1 , D_2 , and D_3 are the inverted values of d_1 , d_2 , and d_3 . Table 3 shows the value mapping table for the 8 different delay times.

delay [µs]	D1	D2	D3	d1	d2	d3	Comments
50	1	1	1	0	0	0	Default value epc700-CSP6-A
5	1	1	0	0	0	1	
10	1	0	1	0	1	0	
20	1	0	0	0	1	1	
100	0	1	1	1	0	0	Default value epc702-CSP6, epc700-QFN16, epc702-QFN16
200	0	1	0	1	0	1	
500	0	0	1	1	1	0	
1000	0	0	0	1	1	1	

Table 3: Delay-time programming table

An un-configured chip is applying a default delay time of e.g. 50µs corresponding to [d₁, d₂, d₃] = 000 and [D₁, D₂, D₃] = 111.

The recovery time is set to o_1 , o_2 , and o_3 , resp. O_1 , O_2 , and O_3 which are the inverted values of o_1 , o_2 , and o_3 . Table 4 shows the value mapping table for the 8 different recovery time delay values.

Recovery time [ms]	O ₁	O ₂	O ₃	O 1	O ₂	O 3	Comments
500	1	1	1	0	0	0	Default value epc700-CSP6-A
10	1	1	0	0	0	1	
20	1	0	1	0	1	0	Default value epc702-CSP6, epc700-QFN16, epc702-QFN16
50	1	0	0	0	1	1	
100	0	1	1	1	0	0	
200	0	1	0	1	0	1	
1000	0	0	1	1	1	0	
1500	0	0	0	1	1	1	

Table 4: Recovery time programming table

An un-configured chip is applying the default recovery time of e.g. 500ms corresponding to $[o_1, o_2, o_3] = 000$ and $[O_1, O_2, O_3] = 111$.

Programming Example

An example for changing the delay time to 100µs is shown in the following diagram:



Figure 9: Programming example to set t_{del} to 100µs



CSP-6 Package

Mechanical Dimensions





no solder mask inside this area -

QFN-16 Package

Note: For sampling only. Limited quantities. Please inquire.





Reflow Solder Profile

For infrared or conventional soldering the solder profile has to follow the recommendations of IPC/JEDEC J-STD-020C (min. revision C) for Pb-free assembly for both types of packages. The peak soldering temperature (T_L) should not exceed +260°C for a maximum of 4 sec.

Packaging Information (all measures in mm)

Tape & Reel Information

The devices are packaged into embossed tapes for automatic placement systems. The tape is wound on 178 mm (7 inch) or 330 mm (13 inch) reels and individually packaged for shipment. General tape-and-reel specification data are available in a separate data sheet and indicate the tape sizes for various package types. Further tape-and-reel specifications can be found in the Electronic Industries Association (EIA) standard 481-1, 481-2, 481-3.



epc does not guarantee that there are no empty cavities in the tape. Thus, the pick-and-place machine should check the presence of a chip during picking.

Ordering Information

Standard products:

Part Number	Part name	Package	RoHS compliance	Packaging Method
P100 009	epc700-CSP6	CSP6	Yes	Reel
P100 039	epc701-CSP6	CSP6	Yes	Reel
P100 059	epc702-CSP6	CSP6	Yes	Reel
P100 060	epc703-CSP6	CSP6	Yes	Reel

Note: The actual revision of the circuit is indicated by an index added to the part name e.g. "-A": epc700-CSP6-A. This is not part of the order information.

QFN package for sampling only. Limited quantities. Please inquire.

Part Number	Part name	Package	RoHS compliance	Packaging Method
P100 038	epc700-QFN16	QFN16	Yes	Reel
P100 040	epc701-QFN16	QFN16	Yes	Reel
P100 057	epc702-QFN16	QFN16	Yes	Reel
P100 058	epc703-QFN16	QFN16	Yes	Reel

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